# A 2-D extension of the sampling algorithm for sparse fourier representations

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Abstract—Recently, a very efficient sampling algorithm for finding a *B*-term fourier representation of given 1-D discrete signal is presented [Gilbert, Guha, Indyk, Muthukrishnan, Strauss; STOC02]. In this paper, we present a modified version of the algorithm, which can be applied to 2-D signals. As in the original algorithm, dependence of the running time on the signal length is polylogarithmic.

#### I. INTRODUCTION

Let

$$\mathbf{x} = (x[0], x[1], \dots, x[n], \dots, x[N-1])$$
(1)

be a complex-valued discrete time signal of length N. The discrete fourier transform (DFT), or the spectrum, of  $\mathbf{x}$ ,

$$\widehat{\mathbf{x}} = (\widehat{x}[0], \widehat{x}[1], \dots, \widehat{x}[k], \dots, \widehat{x}[N-1])$$
(2)

where

$$\widehat{x}[k] := \frac{1}{\sqrt{N}} \sum_{n=0}^{N-1} x[n] e^{-j2\pi nk/N},$$
(3)

is a fundamental tool for analyzing or representing features of the signal **x**. The inverse DFT (IDFT):

$$x[n] = \frac{1}{\sqrt{N}} \sum_{k=0}^{N-1} \widehat{x}[k] e^{j2\pi nk/N},$$
(4)

recovers the signal x completely from the spectrum  $\hat{\mathbf{x}}$ , and these transforms preserve the  $\ell^2$  norm:

$$||\mathbf{x}||_2 = ||\widehat{\mathbf{x}}||_2,\tag{5}$$

where

$$||\mathbf{x}||_{2} := \sqrt{\sum_{n=0}^{N-1} |x[n]|^{2}}.$$
(6)

Let B < N be a positive integer and suppose that the frequency list  $K(B) = (k_1, \ldots, k_B)$  is taken so that  $\hat{x}[k_1], \ldots, \hat{x}[k_B]$  are the *B*-maxima among the spectrum  $\hat{\mathbf{x}}$ , in their absolute values. Then the signal  $\mathbf{r}_{opt(B)}$  defined by

$$r_{\text{opt}(B)}[n] = \frac{1}{\sqrt{N}} \sum_{k \in K(B)} \widehat{x}[k] e^{j2\pi nk/N}$$
(7)

is the best possible  $\ell^2$  approximation of the signal **x** by a *B*-term sum of pure tones  $\psi_k[n] := e^{j2\pi nk/N}/\sqrt{N}$ . We shall refer  $\mathbf{r}_{\text{opt}(B)}$  as the optimal *B*-term (fourier) representation of **x**.

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In those applications that the optimal B-term representation permit a good approximation to the original signal with  $B \ll$ N, the B-term representation, or equivalently, the lists K(B)of frequencies and  $\hat{x}[k_1], \ldots, \hat{x}[k_B]$  of coefficients, would be a nice "digest" of the original signal, in the views of conciseness and accuracy. The computation is obviously feasible, by a combination of the fast fourier transform (FFT) algorithm and some sorting algorithm. The running time has a lower bound  $\Omega(N \log N)$  (here we regard the required precision as a constant), which sounds not so big. However, when we want to make B-term digest for each of x's, in a huge database, each having a really long length N, the cost  $N \log N$  by the number of x's will be so expensive. The formula (3) shows that every fourier coefficient  $\hat{x}[k]$  depends on entire signal x, and it would be natural that one think it were impossible to reduce the computational complexity even with the sparseness condition  $B \ll N$  (say B = 10 independent of N.)

Gilbert, Guha, Indyk, Muthukrishnan, and Strauss [1] have shown a somewhat surprising result that a *near*-optimal *B*-term representation of any signal can be obtained with high probability, within a time whose dependence on N is poly-logarithmic. Let  $\iota$  be a small number such that, if  $||\mathbf{x} - \mathbf{y}||_2^2 \le \iota$  then we consider that two signals  $\mathbf{x}$  and  $\mathbf{y}$  are identical. By a normalization, we assume that  $\iota = 1$  and M is a crude upper bound on  $||x||_2$  after the normalization.

**Theorem 1 ([1]).** There exists an algorithm, on input B < N,  $\epsilon > 0$ , and  $\mathbf{x}$ , with cost  $(B \log(N) \log(M)/\epsilon)^{O(1)}$ , with high probability, output a B-term representation  $\mathbf{r}$  for  $\mathbf{x}$  which satisfies  $||\mathbf{x} - \mathbf{r}||_2 \le (1 + \epsilon)||\mathbf{x} - \mathbf{r}_{opt}(B)||_2$ , where  $\mathbf{r}_{opt}(B)$  is the optimal B-term representation for  $\mathbf{x}$ . The algorithm accesses only  $(B \log(N) \log(M)/\epsilon)^{O(1)}$  samples  $\{x[n] : n \in T\}$  of  $\mathbf{x}$ , where the random sample set T is chosen independently of  $\mathbf{x}$ . (The success probability may be a constant arbitrarily close to 1 and the cost for it is implied in "O(1)".)

Thus, when B is sufficiently smaller than N, the algorithm, which we shall refer as the GGIMS algorithm in the sequel, enables very efficient computation of B-term representations.

Now, suppose that **x** is a signal defined over the 2dimensional domain  $\{0, \ldots, N_1 - 1\} \times \{0, \ldots, N_2 - 1\}$ :

$$\mathbf{x} = (x[n_1, n_2] : 0 \le n_1 < N_1, 0 \le n_2 < N_2)$$
(8)

and let

$$\widehat{\mathbf{x}} = (\widehat{x}[k_1, k_2] : 0 \le k_1 < N_1, 0 \le k_2 < N_2) \tag{9}$$

be its 2-D DFT, where

$$\widehat{x}[k_1, k_2] := \sum_{n_1, n_2} x[n_1, n_2] \psi_{k_1, k_2}[-n_1, -n_2], \qquad (10)$$

and

$$\psi_{k_1,k_2}[n_1,n_2] := \frac{e^{j2\pi \left(\frac{k_1n_1}{N_1} + \frac{k_2n_2}{N_2}\right)}}{\sqrt{N_1N_2}}.$$
(11)

The optimal *B*-term representation for x, i.e. the lists  $K(B) = ((k_{1,1}, k_{2,1}), \ldots, (k_{1,B}, k_{2,B}))$  of frequencies and the corresponding coefficients  $\hat{x}[k_{1,1}, k_{2,1}], \ldots, \hat{x}[k_{1,B}, k_{2,B}]$  where the latter are *B*-maxima of  $\hat{x}$  in absolute value, will be useful in this case also. The goal of this paper is to extend the GGIMS algorithm to an algorithm that efficiently compute near-optimal *B*-term representations for signals defined over a 2-dimensional domain.

The rest of paper is organized as follows: Section II. is a brief description of the GGIMS algorithm [1] for 1-dimensional signals, where we shall discuss what parts of the algorithm are dimension-dependent. In Section III., we modify parts of the GGIMS algorithm so that it can be applied for 2-dimensional signals. Section IV. concludes the paper.

In the sequel, the time or frequency domain  $\{0, 1, ..., N-1\}$  is identified with  $\mathbb{Z}/N\mathbb{Z}$ , the ring of integers modulo N. For simplicity, we assume that  $N, N_1, N_2$  are odd prime numbers.  $\mathbb{E}[X]$  means the expectation of a random variable X.

# II. THE GGIMS ALGORITHM

The GGIMS algorithm composed of 3 parts; Identification, Estimation and Iteration. Given the input signal  $\mathbf{x}$ , the algorithm start with the current list S set to null, and the corresponding at-most-B-term representation  $\mathbf{r}$  is set to 0.

#### A. Identification

Let  $\eta := \epsilon/(2B)$ . The purpose of this part is to output a list  $\Lambda$  of frequencies, whose length is  $2m + 1 = O(1/\eta)$ , which contains (with high probability) all frequencies k satisfying  $|\hat{x}[k]|^2 \ge \eta ||\mathbf{x}||_2^2$ . In this part there exist two procedures; Isolation and Group Testing. The latter procedure call a subroutine that (roughly) estimate the norm of given signal by sampling.

1) Isolation: This procedure constructs 2m + 1 signals  $\mathbf{f}_0, \dots \mathbf{f}_{2m}$  that satisfy (with high probability)

- 1. For each k' such that  $|\widehat{x}[k']|^2 \ge \eta ||\mathbf{x}||_2^2$ , there exists  $i(0 \le i \le 2m+1)$  such that  $\widehat{f}_i[k'] \ge 0.98 ||\mathbf{f}_i||_2^2$ .
- Each f<sub>i</sub> can be sampled by sampling non-adaptively from x in O(m) places.

Here we note that the word "*constructs*" does not mean actual output of the entire signal points  $(f_i[0], f_i[1], \ldots, f_i[N-1])$ .

The procedure only sets up parameters which enable later procedures to sample arbitrary signal points of  $\vec{f_i}$ . To show the construction we define the operator  $R_{\theta,\sigma}$  for  $\sigma, \theta \in \mathbf{Z}/N\mathbf{Z}$ through

$$(R_{\theta,\sigma}f)[n] = e^{j2\pi\theta n/N} f[\sigma n].$$
(12)

Then it holds that

$$(\widehat{R_{\theta,\sigma}}f)[\sigma k + \theta] = \widehat{f}[k].$$
(13)

Also we define the Fejér kernel of length 2m + 1 as

$$H_m[n] := \begin{cases} \frac{\sqrt{N}}{2m+1} & n \in [-m,m] \\ 0 & \text{otherwise.} \end{cases}$$
(14)

It holds that

$$\widehat{H}_m[k] = \begin{cases} \frac{\sin(\pi(2m+1)/N)}{(2m+1)\sin(\pi k/N)} & (k \neq 0) \\ 1 & (k = 0) \end{cases}$$
(15)

and in particular  $\widehat{H}_m[k] \ge 2/\pi$  for  $k \in [-N/(2(2m + 1)), N/(2(2m + 1))]$ . We regard the last interval as "pass region" of  $\mathbf{H}_m$ 

Now, the signals  $\mathbf{f}_0, \ldots \mathbf{f}_{2m}$  are defined as follows: first, pick  $\theta$  and  $\sigma$  at random in  $\mathbf{Z}/N\mathbf{Z}$ , with  $\sigma$  invertible. Then put, for each  $0 \le i \le 2m$ ,

$$\mathbf{f}_i := (e^{j2\pi i n/(2m+1)} \mathbf{H}_m) * (R_{\theta,\sigma} \mathbf{x}), \tag{16}$$

where \* denotes the convolution. In the frequency domain, the spectrum  $\hat{x}$  is first permuted by  $k \mapsto \sigma k + \theta$ , then filtered by "iN/(2m+1)-shifted version of"  $\hat{\mathbf{H}}_m$  to form  $\hat{\mathbf{f}}_i$ .

Using the fact that  $k \mapsto \sigma k + \theta$  is a pair-wise independent permutation (i.e., for all  $k_1 \neq k_2$  and  $k_3 \neq k_4$ ,  $k_1 \mapsto k_3$  and  $k_2 \mapsto k_4$  with probability 1/(N(N-1))), it can be shown that the following holds with high probability: for each k' such that  $|\widehat{x}[k']|^2 \geq \eta ||\mathbf{x}||_2^2$ , there exists  $i(0 \leq i \leq 2m + 1)$  such that  $\widehat{f}_i[k'] \geq 0.98 ||\mathbf{f}_i||_2^2$ . We note that sampling one point from  $\mathbf{f}_i$ can be done by sampling 2m+1 points from  $\mathbf{x}$ , by construction.

2) Group Testing: For each  $\mathbf{f} \in {\{\mathbf{f}_0, \dots, \mathbf{f}_{2m}\}}$ , the procedure determines the frequency k' satisfying  $\widehat{f}[k'] \ge 0.98||\mathbf{f}||_2^2$ , if such k' exists. Then gather such k' from  $\mathbf{f}_i$ 's to form the frequency list  $\Lambda$ . By the construction of  $\mathbf{f}_i$ 's,  $\Lambda$  catches all such frequencies that  $|\widehat{x}[k']|^2 \ge \eta ||\mathbf{x}||_2^2$ .

The procedure utilizes 16 filters

$$\hat{G}_{\ell}[k] = \frac{1}{2} (1 + \cos(\frac{2\pi k}{N} - \frac{2\pi \ell}{16})) \quad (\ell = 0, \dots, 15), \quad (17)$$

each is of length 3. If we define the pass region of  $\hat{\mathbf{G}}_{\ell}$  by

$$pass_{\ell} := \{k : |2\pi k/N - 2\pi \ell/16| \le 2\pi/32\}, \quad (18)$$

then  $\hat{G}_{\ell}[k] \ge 0.99$  for  $k \in \text{pass}_{\ell}$  and these pass regions tile the entire frequency domain. The procedure also uses a subroutine, which will be explained in the next subsection, satisfying the

Fig.	1.	Estimating	Norm
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following: The subroutine makes  $O(\log \log M)$  samples, runs in time polynomial in  $\log(M)$ , on input **f** (of length N) with biggest fourier coefficient  $\widehat{f}[k]$ , it returns, with high probability, a random output X such that:

1.  $X \leq ||\mathbf{f}||_2^2, \forall \mathbf{f}.$ 2. If  $|\hat{f}[k]|^2 \geq 0.95 ||\mathbf{f}||_2^2$  then  $X \geq 0.5 ||\mathbf{f}||_2^2$ .

Now assume that we are seeking unknown k' which have at least 98% energy of f. Without loss of generality, we assume that  $k' \in \text{pass}_0$ . For each  $\ell = 0, 1, \dots, 15$ , We estimate the norm  $||\mathbf{f} * G_{\ell}||$  using the subroutine. For  $\ell = 0$ , it can be shown that  $|\mathbf{f} * \mathbf{G}_0|^2 \ge 0.95 ||\mathbf{f} * \mathbf{G}_0||_2^2$  by combining the facts  $k' \in \text{pass}_0$ ,  $\hat{G}_0[k'] \ge 0.99$  and  $\hat{f}[k'] \ge 0.98 ||\mathbf{f}||_2^2$ . Then the second property of the subroutine, the returned value X satisfies  $X \ge 0.5 ||\mathbf{f} * \hat{G}_0||_2^2$ . It also can be shown that  $0.5 ||\mathbf{f} * \hat{G}_0||_2^2 \ge 0.48 ||\mathbf{f}||_2^2$ . On the other hand, for  $\ell = 4$ , it can be shown that the returned value X of the subroutine satisfies  $X \leq ||\mathbf{f} * \hat{G}_4||_2^2 \leq 0.38 ||\mathbf{f}||_2^2$ , from the facts that k' is far from pass<sub>4</sub> and that the contributions to  $||\mathbf{f} * \hat{G}_4||_2^2$ from frequencies other than k' is small (at most  $0.02||\mathbf{f}||_2^2$ ). For  $\ell = 5, 6, \dots, 12$ , we obtain  $X \leq 0.38 ||\mathbf{f}||_2^2$  in the same way. So, we can eliminate the possibilities that  $k' \in pass_{\ell}$  for  $\ell = 4, \ldots, 12$ . In this manner, we can always eliminate 9/16possibilities. The remaining region forms a cyclic interval of lentgth at most 7N/16. By applying the operator  $R_{0,2}$ , we can dilate the remaining frequency region by two. Then applying 16 filters, we can halve the possibilities again. Repeating that  $O(\log(N))$  times, we learn k'.

3) Estimating Norm: The subroutine for estimating norm is shown in Fig. 1, where  $\beta > 0$  is a small constant (independent of N and M), and the number r of samples is  $O(\log \log M)$ , and in fact we can reuse the same samples in the loop.  $K_c f$  stands means "f clipped at the ceiling c", i.e,

$$(K_c f)[n] := \begin{cases} f[n] & (|f[n]| \le c/\sqrt{N}) \\ 0 & \text{otherwise.} \end{cases}$$
(19)

Intuition behind the algorithm is as follows: Let  $Y = \frac{1}{r} \sum_{i=1}^{r} N|f[n_i]|^2$ . Then  $\mathsf{E}[Y] = ||\mathbf{f}||_2^2$  and it would be a good estimate if the variance of Y were small. The variance comes from spikes, so we use the clipped version X instead. It is satisfied that  $0 \le \mathsf{E}[X] \le \mathsf{E}[Y] = ||\mathbf{f}||_2^2$  and X has small variance.

If 95% of **f** is concentrated in a pure frequency, then the energy of spikes are small, so  $E[X] \approx E[Y] = ||\mathbf{f}||_2^2$ , and since variance of X is small, X is a good estimation. The algorithm sets a ceiling c, above which all values of the function are clipped as if they were spikes. The algorithm gradually lowers the clipping celing c until the energy estimate is consistent with the ceiling value.

## B. Estimation

This part estimates  $\hat{x}[k]$ , for each  $k \in \Lambda$ , by a sampling algorithm.

The sampling algorithm used here takes inputs  $(\mathbf{a}, k, \mu)$ , where **a** is a signal of length N, k a frequency,  $\mu$  an accuracy parameter. Then the algorithm runs with cost  $O(\log(M)/\mu)$ , making  $O(1/\mu)$  samplings from **a**, and returns the estimation  $\tilde{a}[k]$ , which satisfies with high probability that  $|\tilde{a}[k] - \hat{a}[k]|^2 \le \epsilon ||\mathbf{a}||_2^2$ . The estimation  $\tilde{a}[k]$  is computed as follows: let  $t = O(1/\mu)$  and pick t random positions  $n_1, \ldots, n_t$ . Put  $\mathbf{a}'' = N/t \sum_{h=1}^t \Delta_{n_h}$ , where  $\Delta_n$  is the delta function. Then compute X = a''[k] and return it as the estimate. It can be shown that  $\mathsf{E}[X] = \hat{a}[k]$  and  $\mathsf{E}[|X - \hat{a}[k]|^2] \le O(\epsilon ||\mathbf{a}||_2^2)$ .

For the estimation of  $\hat{x}[k]$ , an accuracy of  $\mu = 1/poly(B\log(N)\log(M/\delta)/\epsilon)$  (where  $\delta = ||\mathbf{x} - \mathbf{r}_{opt(B)}||_2$ ) is required.

# C. Iteration

Now we have the estimated fourier coefficients  $\tilde{x}[k]$  for  $k \in \Lambda$ . Let  $\tilde{x}[k']$  be the maximum (in abosolute value) among them. Add k' to the current coefficients list S, and update the current representation  $\mathbf{r}$  by  $\mathbf{r} \leftarrow \mathbf{r} + \tilde{x}[k']\psi_{k'}$ . Then set  $\mathbf{x} \leftarrow \mathbf{x} - \tilde{x}[k']\psi_{k'}$  and A.. However, we should note that the update of  $\mathbf{x}$  just before goto is in fact impossible, since we can access  $\mathbf{x}$  only via sampling. Actually, after 2nd iteration, we sample from  $\mathbf{x} - \mathbf{r}$  whenever we need sample from  $\mathbf{x}$ , using the maintained representation  $\mathbf{r}$ .

The algorithm may or may not halt just after that the representation **r** has grown to *B*-terms. Here we do not mention the accurate halting condition, however, it is guaranteed to halt within  $poly(B \log(N) \log(M/\delta)/\epsilon)$  iterations.

#### **III.** MODIFICATIONS FOR 2-D SIGNALS

The strategy of GGIMS algorithm is not dimensiondependent, However, two key steps in the Identification part needs a modification for 2-D signals. Here we describe modifications for Isolation step and Group Testing step. In the sequel, signals are defined over the 2-dimensional domain  $\mathbf{Z}/N_1\mathbf{Z} \times \mathbf{Z}/N_2\mathbf{Z}$ .

# A. Identification (2-D)

1) Isolation (2-D): Let  $\eta := \epsilon/(2B)$ .We will construct  $(2m_1 + 1) \times (2m_2 + 1)$  signals  $f_{i_1,i_2}[n_1,n_2](i_1 =$ 

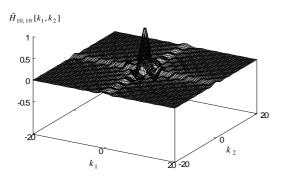


Fig. 2.  $\hat{H}_{10,10}[k_1, k_2](N_1 = N_2 = 40)$ 

 $0, 1, \dots, 2m_1, i_2 = 0, 1, \dots, 2m_2$ , where  $(2m_1+1) \times (2m_2+1) = O(1/\eta)$ .

We define the operator  $R_{\theta_1,\sigma_1,\theta_2,\sigma_2}$  through

$$R_{\theta_1,\sigma_1,\theta_2,\sigma_2} x[n_1,n_2] = e^{j2\pi(\theta_1 n_1/N_1 + \theta_2 n_2/N_2)} x[\sigma_1 n_1,\sigma_2 n_2] \quad (20)$$

and 2-D Fejér kernel by

$$H_{m_1,m_2}[n_1,n_2] = \begin{cases} \frac{\sqrt{N_1N_2}}{(2m_1+1)(2m_2+1)} & (n_1,n_2) \in \\ & [-m_1,m_1] \times [-m_2,m_2] \\ 0 & (otherwise) \end{cases}$$
(21)

See Fig.2 for the spectrum.

Then the signals  $f_{i_1,i_2}[n_1,n_2](i_1 = 0, 1, \ldots, 2m_1, i_2 = 0, 1, \ldots, 2m_2)$  are constructed as follows: pick  $\sigma_1, \theta_1 \in \mathbf{Z}/N_1\mathbf{Z}$  at random, with  $\sigma_1$  invertible. Independently, pick  $\sigma_2, \theta_2 \in \mathbf{Z}/N_2\mathbf{Z}$  at random, with  $\sigma_2$  invertible. Then set

$$f_{i_1,i_2}[n_1, n_2] := (e^{j2\pi(i_1n_1/(2m_1+1)+i_2n_2/(2m_2+1))}H_{m_1,m_2}[n_1, n_2]) * R_{\theta_1,\sigma_1,\theta_2,\sigma_2}x[n_1, n_2].$$
(22)

It can be shown that they have the following (expected) properties:

- 1. For each  $k'_1, k'_2$  such that  $|\hat{x}[k'_1, k'_2]|^2 \ge \eta ||\mathbf{x}||_2^2$ , there exists  $(i_1, i_2)(0 \le i_1 \le 2m_1 + 1, 0 \le i_2 \le 2m_2 + 1)$  such that  $\hat{f}_{i_1, i_2}[k'_1, k'_2] \le 0.98 ||\mathbf{f}_{i_1, i_2}||_2^2$ .
- 2. Each  $\mathbf{f}_{i_1,i_2}$  can be sampled by sampling non-adaptively from  $\mathbf{x}$  in  $O(m_1m_2)$  places.

2) *Group Testing* (2-*D*): In 2-D case, we rely on the following 256 filters (see Fig. 3):

$$\hat{G}_{\ell_1,\ell_2}[k_1,k_2] = \frac{1}{4} \left\{ 1 + \cos\left(\frac{2\pi k_1}{N_1} - \frac{2\pi \ell_1}{16}\right) \right\}$$
$$\cdot \left\{ 1 + \cos\left(\frac{2\pi k_2}{N_2} - \frac{2\pi \ell_2}{16}\right) \right\}$$
$$(0 \le \ell_1 < 16, 0 \le \ell_2 < 16) \quad (23)$$

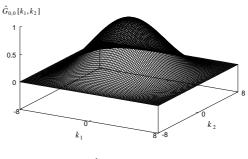


Fig. 3. The filter  $\hat{G}_{0,0}[k_1, k_2]$   $(N_1 = N_2 = 16)$ 

In the time domain,  $\mathbf{G}_{\ell_1,\ell_2}$  has a support of  $3 \times 3$  square.

With these filters, we can eliminate  $1 - (7/16)^2 = 207/256$ (instead of 1 - 7/16 in 1-D case) of possibilities in one step of Group Testing.

#### B. Complexity of 2-D algorithm

By the above modifications of the algorithm, we obtain a natural 2-D extension of Theorem 1:

**Theorem 2.** There exists an algorithm, on input  $B < N_1N_2$ ,  $\epsilon > 0$ , and a signal **x** over the 2-dimensional domain  $\{0, \ldots, N_1 - 1\} \times \{0, \ldots, N_2 - 1\}$ , with cost  $(B \log(N_1) \log(N_2) \log(M)/\epsilon)^{O(1)}$ , with high probability, output a B-term representation **r** for **x** which satisfies  $||\mathbf{x} - \mathbf{r}||_2 \leq (1 + \epsilon)||\mathbf{x} - \mathbf{r}_{opt}(B)||_2$ , where  $\mathbf{r}_{opt}(B)$  is the optimal B-term representation for **x**. The algorithm accesses only  $(B \log(N_1) \log(N_2) \log(M)/\epsilon)^{O(1)}$  samples  $\{x[n_1, n_2] : (n_1, n_2) \in T\}$  of **x**, where the random sample set T is chosen independently of **x**. (The success probability may be a constant arbitrarily close to 1 and the cost for it is implied in "O(1)".)

## IV. CONCLUSION

In this paper, we have presented an efficient algorithm that finds a near-optimal *B*-term fourier representation of 2-D signals, by modifying a part of GGIMS algorithm [1].

The running time of the algorithm depends on signal amount only polylogarithmically, so the algorithm will be useful for digesting vast 2-D signals.

The algorithm has many hidden constants that would affect practical use. So more fine analysis of the algorithm is still needed.

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